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2.161 Signal Processing: Continuous and Discrete  
Fall 2008

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MASSACHUSETTS INSTITUTE OF TECHNOLOGY  
DEPARTMENT OF MECHANICAL ENGINEERING

2.161 Signal Processing - Continuous and Discrete  
Fall Term 2008

**Problem Set 4**

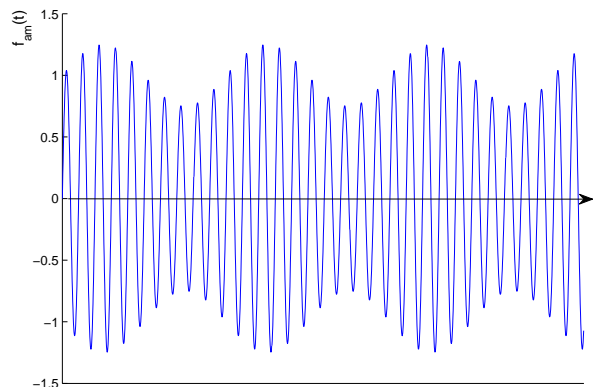
**Assigned:** Oct. 2, 2008

**Due:** Oct. 9, 2008

**Problem 1:** An AM (amplitude-modulated) radio signal  $f_{AM}(t)$  is described by

$$f_{AM}(t) = (1 + af_{audio}(t)) \sin(\Omega_c t)$$

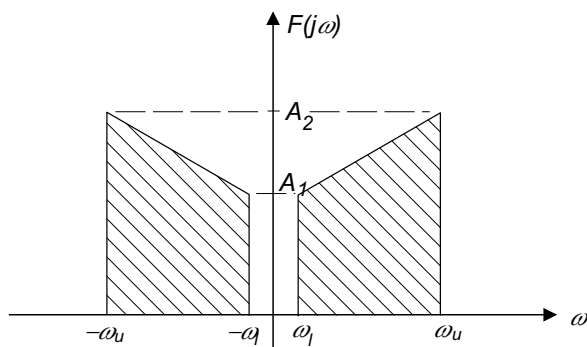
where  $f_{audio}(t)$  is the audio signal,  $\sin(\Omega_c t)$  is known as radio-frequency *carrier* signal ( $f_c = 500 - 1600$  kHz - the AM band), and  $a$  is a positive constant that determines the *modulation depth*. (Note that we require  $|af_{audio}(t)| < 1$  otherwise we have over-modulation.) the following figure shows an AM signal with an “audio” waveform that is a simple low frequency sinusoid. You can see how the audio signal “modulates” the amplitude of the rf signal.



- (a) Sketch the magnitude of the Fourier transform of  $f_{AM}(t)$  when  $f_{audio}(t) = 0$ .
- (b) Let  $a = 0.5$ , and sketch the magnitude of the Fourier transform of  $f_{AM}(t)$  when

$$f_{audio}(t) = 0.5 \cos(2\pi \cdot 1000t) + 0.25 \cos(2\pi \cdot 2000t)$$

- (**Hint:** There is no need to actually compute the FT. Consider expanding  $f_{AM}(t)$ , or simply use properties of the FT.)
- (c) Use your result from (b) to generalize, and sketch the magnitude spectrum of  $f_{AM}(t)$  when  $f_{audio}(t)$  has a spectrum (again let  $a = 0.5$ ):



- (d) If  $f_{audio}$  has a bandwidth  $B = \Omega_u - \Omega_l$ , what is the bandwidth of a band-pass filter that would be necessary to select the signal  $f_{AM}(t)$  out of all the other AM radio stations?

**Problem 2:** We generally ignore in the phase response in filter design. Although you might wish for a “zero-phase” filter, you can see from the class handout on causality that a filter with a purely real frequency response is acausal, and as such cannot be implemented in a physical system. The following are a pair of tricks that may be used to do “off-line” zero-phase filtering of recorded data. (Note: these methods are used frequently in digital signal processing - it is difficult to do this in continuous time.)

Assume that you have a filter  $H(j\Omega)$  with arbitrary phase response  $\angle H(j\Omega)$ , and that your input signal is  $f(t)$  is recorded on a tape-recorder that can be played forwards or backwards.

- Method (1)**
1. Pass  $f(t)$  through the filter and record the output  $g(t)$  on another tape recorder.
  2. Play  $g(t)$  backwards through the filter (that is the filter input is  $g(-t)$ ) and record the output  $x(t)$ .
  3. The filtered output is found by playing the  $x(t)$  backwards, that is  $y(t) = x(-t)$ .

- Method (2)**
1. Pass  $f(t)$  through the filter and record the output  $g(t)$ .
  2. Reverse  $f(t)$  so as to pass  $f(-t)$  through the filter and generate  $x(t)$ .
  3. Reverse  $x(t)$  and sum with  $g(t)$  to form the output  $y(t) = g(t) + x(-t)$ .

Show that both methods generate an overall filter that has no phase shift, and find the overall magnitude response  $|H_{eq}(j\Omega)|$  in each case. Hint:  $\mathcal{F}\{f(-t)\} = \bar{F}(j\Omega)$ .

**Problem 3:** Problem 5 in Problem Set 2 examined an *all-pass* filter with a transfer function

$$H(s) = \frac{s - a}{s + a} \quad a > 0$$

and you showed that this filter had a frequency response in which  $|H(j\Omega)| = 1$  at all frequencies.

Design an op-amp based first-order all-pass filter of this form that will have a phase shift of  $-90^\circ$  at a frequency of 50 Hz. (Consider a modified form of the 3 op-amp circuit described in the handout - noting that you only need a first-order system.) Find “appropriate” values

for all resistors and capacitors.

**Problem 4:** Consider the second-order bandpass filter with transfer function

$$H_{bp}(s) = \frac{a_1 s}{s^2 + a_1 s + a_0}.$$

Many books on signal processing express this transfer function in terms of two parameters  $\Omega_p$  and  $Q$ ,

$$H_{bp}(s) = \frac{\frac{\Omega_p}{Q} s}{s^2 + \frac{\Omega_p}{Q} s + \Omega_p^2}$$

where  $\Omega_p$  is the (approximate) peak frequency (center of the passband), and  $Q$  is known as the “quality” factor.

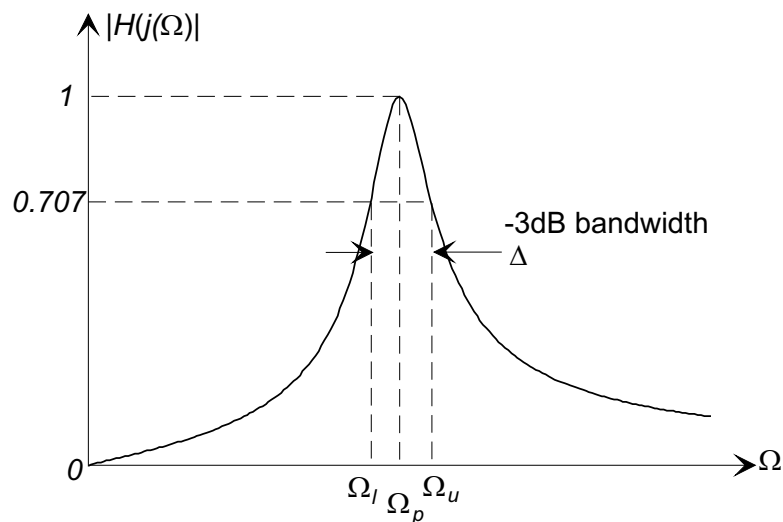
**Aside:** If you compare this to the classic second-order system description used in system dynamics and control, that is

$$H_{bp}(s) = \frac{2\zeta\Omega_n s}{s^2 + 2\zeta\Omega_n s + \Omega_n^2}$$

where  $\Omega_n$  is the undamped natural frequency, and  $\zeta$  is the damping ratio, you can see that

$$Q = \frac{1}{2\zeta}.$$

In this problem we examine the relationship between  $Q$  and the -3dB bandwidth of the second-order filter. Consider the magnitude plot below:



Let  $\Omega_u$  and  $\Omega_l$  be the upper and lower -3db (0.707) response frequencies as shown, and let  $\Delta = \Omega_u - \Omega_l$  be the -3dB bandwidth.

(a) Show

$$\begin{aligned}\Omega_u &= \Omega_p \sqrt{1 + \frac{1}{2Q^2} + \frac{1}{Q} \sqrt{1 + \left(\frac{1}{2Q}\right)^2}} \\ \Omega_l &= \Omega_p \sqrt{1 + \frac{1}{2Q^2} - \frac{1}{Q} \sqrt{1 + \left(\frac{1}{2Q}\right)^2}}\end{aligned}$$

(b) Use these results to show that the -3dB bandwidth of the second-order filter is

$$\Delta = \frac{\Omega_p}{Q}$$

**Hint:** Write  $\Delta = \sqrt{a+b} - \sqrt{a-b}$ .

(c) Determine the transfer function of a second-order bandpass filter with a center frequency of 100 Hz., and a -3dB bandwidth of 10 Hz.

**Problem 5:** A sampling system takes samples at regular intervals  $\Delta T$ . Assume we have a sinusoid

$$y(t) = \sin \Omega t,$$

so that the samples are  $y(n\Delta T) = \sin n\Omega\Delta T$ . We know that if the frequency  $\Omega$  is greater than the Nyquist frequency  $\Omega_N = \pi/\Delta T$ , the sample set is *aliased*.

Now consider two sinusoids, one  $y(t) = \sin n\Omega_0\Delta T$  with a frequency  $\Omega_0$  that is below the Nyquist frequency, and another with frequency  $\Omega_1$  **above** the Nyquist frequency.

(a) Assume  $y_1(t) = \sin \Omega_1 t$  where

$$\Omega_1 = 2k\Omega_N - \Omega_0, \quad k = 1 \dots \infty \text{ is any positive integer.}$$

Show that the sample sets are related by  $y_1(n\Delta T) = -y(n\Delta T) = -\sin(n\Omega_0\Delta T)$ ,

(b) Repeat part (a) with

$$\Omega_1 = 2k\Omega_N + \Omega_0, \quad k = 1 \dots \infty \text{ is any positive integer.}$$

and show that in this case the sample sets are identical, that is  $y_2(n\Delta T) = y(n\Delta T) = \sin(n\Omega_0\Delta T)$ .

(c) Use the results of (a) and (b) to graphically demonstrate the concept of “frequency folding” of aliased sinusoids.

(d) A periodic waveform is written as a Fourier Series

$$y(t) = 5 \sin(2\pi(25)t) + 2 \sin(2\pi(75)t) + 3 \sin(2\pi(125)t).$$

If the waveform is sampled at 100 samples/sec, determine the frequencies and amplitudes of the spectral components in the sampled waveform. (Hint: The results of parts (a) and (b) should help.)